



# Utility Grid Interfaced Solar Water Pumping System Using PMSM Grid with Improved Power Quality

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## ABSTRACT

In this paper a PMSM water pumping system is proposed which is operated from PVA and utility grid in parallel. During no solar irradiation or low irradiation the utility grid supports the load connected to it. The PMSM is operated using sensor less field oriented control with speed estimator. The FOC scheme takes voltage and current feedback from the motor generating estimated speed of the machine using MRAS controller. The MRAS controller also generated estimated rotor angle theta for the operation of Parks and Clarks transformations. The FOC scheme is updated with DSM-PI controller which generates stable speed for PMSM rather compared to conventional PI controller. The model is developed in MATLAB software with comparative results.

**Keywords:** SWP PMSM; Sensor-less vector control; Utility grid; DSM –PI Controller, Sliding mode observer.

## INTRODUCTION

Rising energy concerns have motivated the researchers to search for alternate energy solutions for reducing the utilization of conventional energy sources. Recently various applications have been identified where renewable energy sources (RES) can replace the use of conventional energy sources.

Water pumping is one of the potential areas utilizing large amount of fossil fuels and therefore, the utilization of RES for this application would reduce the emission of greenhouse gases and reduce the carbon foot print [1].

Due to its modular structure, declining installation cost, and zero operating cost, solar photovoltaic (PV) based power generation is gaining wider acceptability [2]. Although solar PV integrated water pumping system (WPS) presents a feasible solution, however, intermittent nature of solar energy limits its use for active hours only i.e. when solar insolation is available. For an effective utilization of WPS, this drawback needs to be resolved. Some of the solutions to this problem are connecting the battery energy storage at the DC link, use of pump storage and fuel cell as a storage medium. However, these solutions also have their flaws such as increased system complexity, cost and space requirement [3]. As the grid is an infinite source of energy, some of the existing work suggests the integration of utility grid to SWP system. An integration to utility grid enhances the system utilization and improves the system reliability. Although some the existing literature focus on power factor correction, however, none of the existing work focuses on problems associated with the weak grid. As most of the water pumps are located at the radial end of



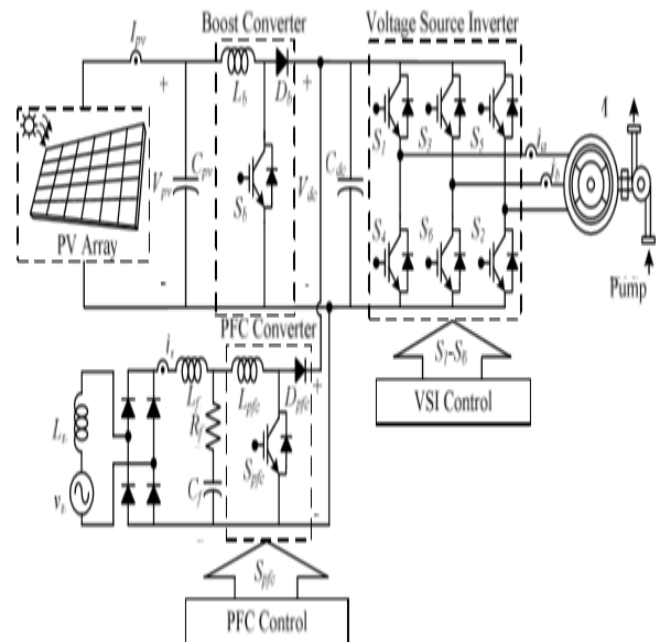
the utility grid, these systems are exposed to various power quality (PQ) problems such as voltage sag, voltage swell, grid current distortion and grid voltage distortion. For the proper operation of SWP system, these issues need to be resolved. This work aims at mitigation of some of the above mentioned issues for proper operation of Water pumping system.

The salient features of presented work are highlighted as follows.

1. The solar water pumping system is integrated to the single phase utility grid to facilitate an uninterrupted water pumping independent of available solar insolation.
2. A double second order generalized integrator quadrature signal generator (DSOGI-QSG) control structure is implemented for filtering the distorted grid voltage and to extract its fundamental component.
3. The PQ issues such as voltage sag, voltage swell, grid current distortion and grid voltage distortion, are mitigated so that the presented system follows the IEEE standard for PQ in grid connected system.
4. To improve the system reliability and to reduce the system cost, a sensor-less vector control technique is implemented for controlling the speed of the PMSM.

## II SYSTEM ARCHITECTURE AND CONTROL

The system configuration of grid integrated solar water pumping system using PMSM drive is shown in Fig 2.1. This system comprises of a solar PV array, a boost converter for maximum power point tracking (MPPT), a voltage source inverter to drive the motor, another boost converter for power factor correction (PFC) and a PMSM coupled to pump. The speed of the PMSM is regulated using sensor-less vector control technique.



**Fig. 2.1 Proposed system configuration**

An incremental conductance (INC) algorithm is utilized for MPPT through change in duty ratio of switch  $S_b$ . The grid is feeding the DC link using PFC converter. The UPF is maintained at AC mains through proper switching of switch  $S_{pfc}$ . A RC filter ( $R_f$ ,  $C_f$ ) and an interfacing inductor ( $L_f$ ) are used for removing switching harmonics and current ripples, respectively.

### 2.2 SYSTEM CONTROL

The control of proposed Water pumping system is divided into three parts. First one is MPPT control using INC MPPT algorithm. Second one is speed control of PMSM using sensor-less vector control technique as shown in Fig. 2.2. A back electromotive force (EMF) based technique is used for speed and position estimation. Third one is power flow control from utility grid. This is achieved using unit vector template (UVT) technique. For proper operation even during distorted grid voltage conditions, first the grid voltage is filtered using DSOGI-QSG as shown in Fig. 2.3. The DSOGI-QSG removes the harmonic component and DC offset.



The transfer function for DSOGI-QSG is written as,

$$D_{DSOGI} = \frac{v_f(s)}{v_s(s)} = \frac{c_1 c_2 \omega^2 s^2}{(s^2 + c_2 \omega s + \omega^2)(s^2 + \omega^2) + c_1 c_2 \omega^2 s^2}$$

$$Q_{DSOGI} = \frac{qv_f(s)}{v_s(s)} = \frac{c_1 c_2 \omega^3 s}{(s^2 + c_2 \omega s + \omega^2)(s^2 + \omega^2) + c_1 c_2 \omega^2 s^2}$$

Equation no.2.1

The filtered voltage is divided with the peak voltage for the generation of unit template. The unit template is multiplied with the voltage regulator output for the generation of reference current. The sensed and reference currents are compared for generating the gating pulses for Spfc.

The sensor less control technique implements the FOC algorithm by estimating the position of the motor without using position sensors. Figure 2.3 is a simplified block diagram of the position estimator function. Motor position and speed are estimated based on measured currents and calculated voltages.

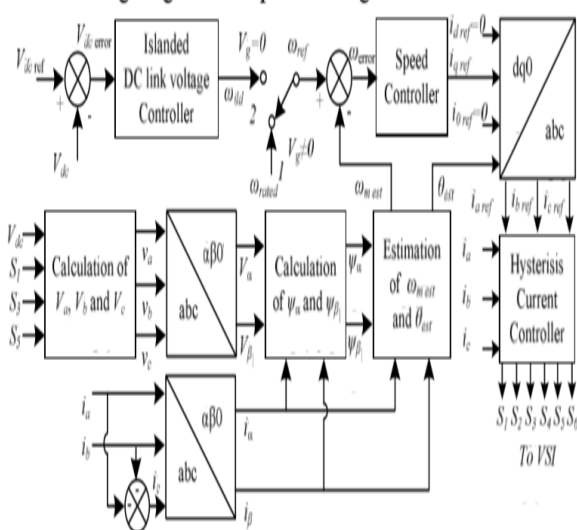


Fig. 2.2 Sensor-less field oriented control scheme of PMSM

### 2.3 PI and Dual sliding mode PI controller

In a normal PI controller the gains are fixed at one particular value. Distinctly in DSM-PI controller gain values are constantly changed according to the error generated. DSM-PI controller is used to derive current signal from speed error input. This DSM-PI controller continuously monitors and gains of proportional and integral gains respectively and consequently the response time is minimized. The main advantage of this scheme is settling time of the speed-time response is reduced with reduction in oscillations and disturbances.

A PI controller has proportional and integral gains fixed at one particular value, which remains constant even for higher value or lower value of error. Whereas in DSM-PI

Controller the values of gains and are variable w.r.t. the error generated. If the speed error is high, value of gains and are increased and if it's low the gain values are decreased so as to reduce the settling time.

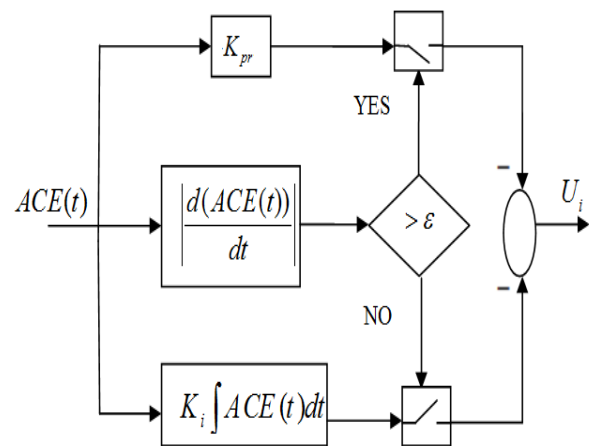
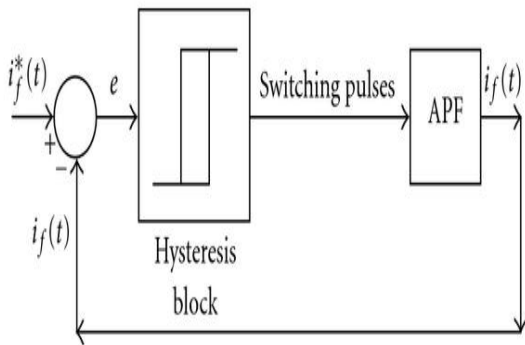


Fig. 2.3 Dual PI Controller

### 2.4 Hysteresis current control Method:

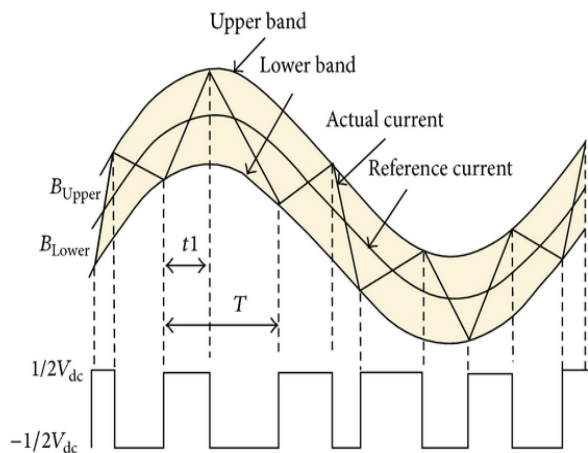
Hysteresis current control is a method of generating the required triggering pulses by comparing the error signal with that of the hysteresis band and it is used for controlling the voltage source inverter so that the output current is generated from the filter will follow the reference current waveform is shown in Figure 2.4



**Figure.2.4 Hysteresis Control**

This method controls the switches of the voltage source inverter asynchronously to ramp the current through the inductor up and down, so that it follows the reference current. Hysteresis current control is the easiest control method to implement in the real time.

Figure 2.5 illustrates the ramping of the current between the two limits where the upper hysteresis limit is the sum of the reference current and the maximum error or the difference between the upper limit and the reference current and for the lower hysteresis limit, it is the subtraction of the reference current and the minimum error. Supposing the value for the minimum and maximum error should be the same. As a result, the hysteresis bandwidth is equal to two times of error .



**Figure. 2.5 Hysteresis Band**

**CHAPTER-III  
SIMULATION RESULT AND DISCUSSION**

The complete design related to the project is created in Matlab& Simulation using Sim Power System Toolbox and thereby analysis the different MPPT Technique. .This designing is conducted in two stage stages:-

- 1 .Grid interfaced PV-INC MPPT system with PI Controller Using for PMSM Motor Drive.
2. Grid interfaced PV-INC MPPT system with Dual sliding mode- PI Controller Using for PMSM Motor Drive.

**3.1 Simulation model parameter:**

| Parameter                 | Value           |
|---------------------------|-----------------|
| Irradiation               | 1000            |
| Temperature               | 30              |
| Voltage of PV Array       | 300V            |
| Current of PV Array       | 8 Amp           |
| Boost Voltage of PV Array | 600V            |
| Inductor(L)               | 1milli henney   |
| Capacitor (Cin)           | 47 micro farad  |
| Capacitor (Cout)          | 2200micro farad |
| PWM Switching Frequency   | 10000           |

**3.2: Proposed test system with PVA connected to grid**

Grid interfaced PV-INC MPPT system with PI Controller Using for PMSM Motor Drive is shown in fig. no. 3.1. The PFC circuit connected to grid and PVA for higher voltage generation with low ripple. The PVA uses incremental conductance method as MPPT for control of the PFC converter connected to it. The grid connected PFC converter uses voltage-oriented control for the control of output voltage. The below are the MPPT method and voltage-oriented control models.

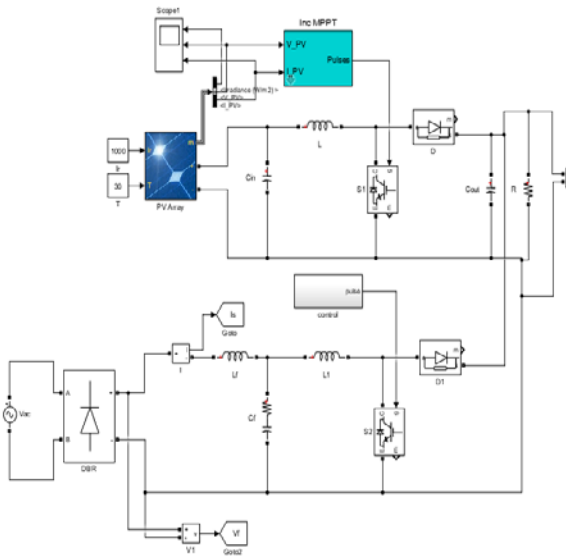


Fig. : 3.1 proposed test system with PVA connected to grid

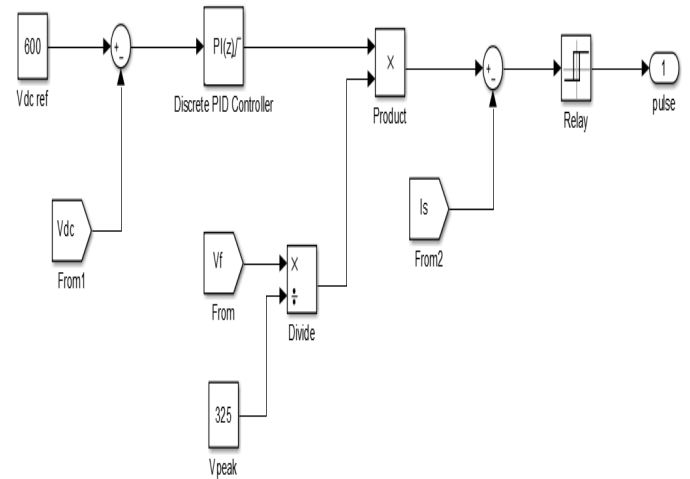


Fig. : 3.3 DC-DC booster converter controller of the grid connection

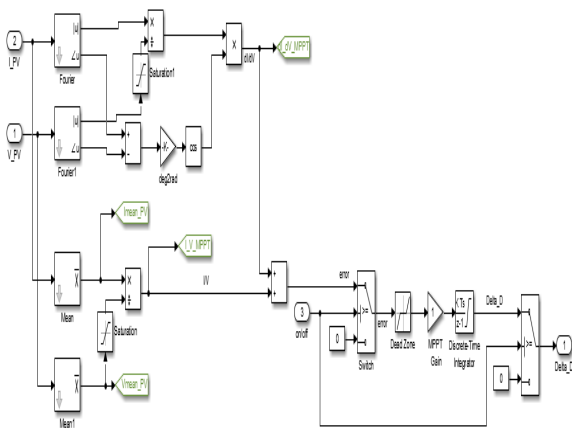


Fig.: 3.2 Incremental conductance MPPT method modeling

In the shown in fig.3.3 controller the reference DC voltage taken is 600V which is maximum value of 440Vrms line to line voltage. The 600V is given as input to the sensor less FOC of PMSM drive for water pumping operation. The below is the modelling of water pumping system with PMSM drive using sensor less FOC

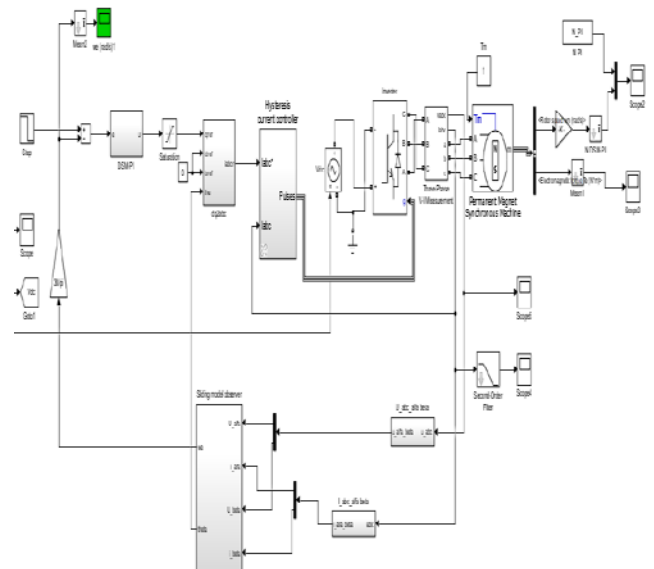


Fig.3.4 FOC scheme with MRAS control for speed control of PMSM

The shown in figure 3.5 the reference speed (1500rpm), estimated speed and measured speed comparison.

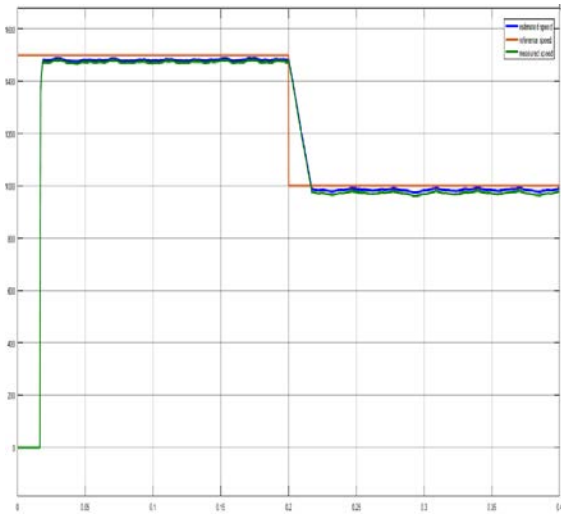


Fig. 3.5 Speed of the machine compared with reference value.

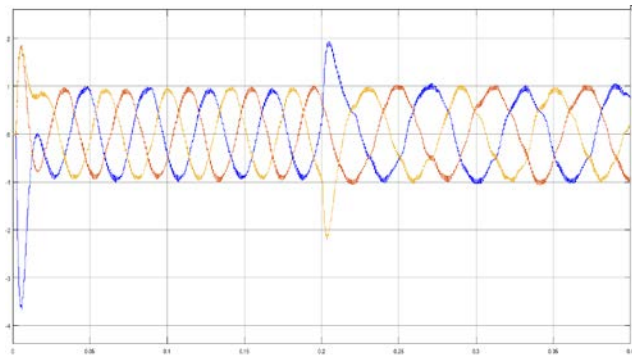


Fig. 3.6 Voltages and currents of PMSM

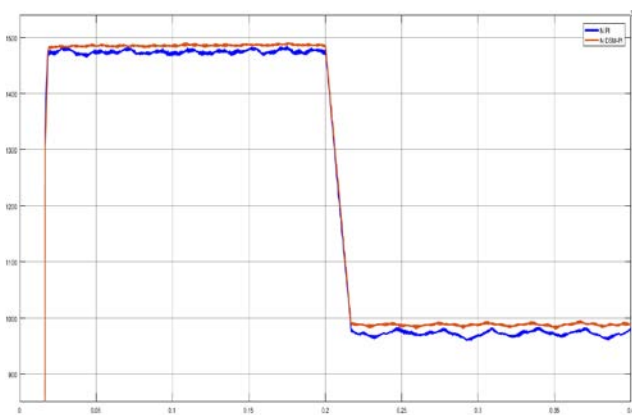


Fig. 3.7 Speed comparison of PMSM with PI and DSM-PI controller

### 3.2 HARMONICS ANALYSIS BETWEEN PI-CONTROLLER & DSM-PI CONTROLLER

The comparison the THD of the DSM-PI controller is less as compared to PI controller during 1500rpm reference value. This is shown in figure 3.8 and 3.9.

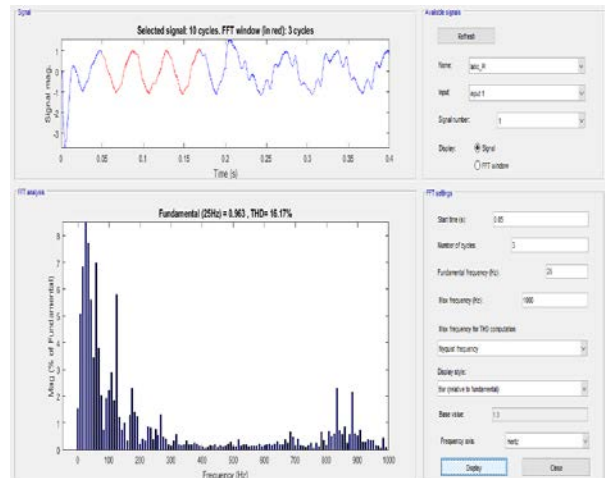


Fig. 3.8 THD analysis of PMSM stator current with PI controller

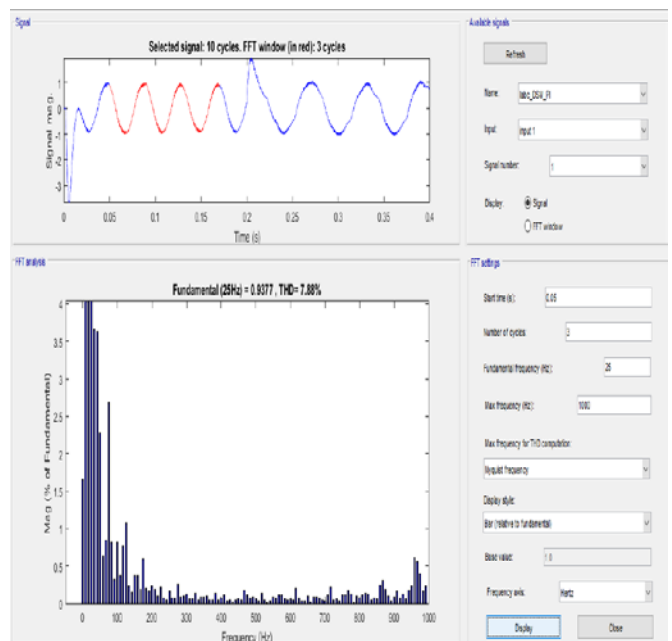


Fig. 3.9 THD analysis of PMSM stator current with DSM-PI controller



## CONCLUSION & FUTURE SCOPE

As per the graphs generated with comparison of PI and DSM-PI controller of the FOC structure operating PMSM it can be considered that the test system with DSM-PI controller is more stable. The speed comparison denotes that measured value of the DSM-PI controller has less disturbance and precise value generation. The THD of the FOC system with PI controller is recorded at 16.17% where the DSM-PI controller is recorded at 7.88% at 1500rpm reference. However the power sharing from PVA and grid is constant with respect to change in solar irradiation in both the models. The FOC controller can further be updated by optimization controller for better performance of the machine with exact speed generation and with least ripple. The THD can be further reduced below 5% with more stabilized control system with faster response rate. The inverter can be replaced with multi-level inverter topologies for reduction of THD in the stator current of the machine.

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